Wavelets: A family of orthobases for L2(1R)

Wavelets are a class of orthobases that represent a signal using the differences between nested approximations at different scales.

It is best to first look at a couple of examples, and then generalize.

Example 1: Shannon Wavelets

BT = { Signals bandlimited to TT }

We know that if

$$\theta_{o}(t) = \frac{\sin(\pi t)}{\pi t}$$

then {\mathbb{G}_0(+-K)} kez is an \(\precedel{\precedel}\) L-basis for B_TT.

Recall that given an arbitrary X(+)

 $\langle X(t), \mathcal{O}(t-k) \rangle = lowpass filter X(t), then sample at t=N$

55

The closest bandlimited signal to X(+) can be written as $\hat{\chi}(t) = \sum_{k=-\infty}^{\infty} \langle \chi(t), \theta_{o}(t-k) \rangle \theta_{o}(t-k)$ we will write this as $\hat{x}(t) = P_{n_{\pi}}[x(t)]$ Projection on to the space BT Now suppose we compress (in time) the sincs by a factor of 2. Set $S_1(+) = \sqrt{2} \cdot S_0(2+)$ Chosen to make $||S_1||_{L_2} = 1$ = $\sqrt{2}$ sin $(2\pi t)$ $= \sqrt{\frac{1}{2}} \cdot \frac{\sin(2\pi\tau)}{2\pi\tau}$ We have seen before that the set $\{\emptyset, (+-k/2)\}_{k\in\mathbb{Z}} = \{\sqrt{2}, \emptyset, (2+-k)\}_{k=-\infty}^{\infty}$ [(sincs w/ bandwith 211 at half-integer shifts) is an orthonormal basis for BZTT.

Similarly, if we define
$$\theta_2(t) = \sqrt{2} \theta_1(2t) = 2\theta_0(4t)$$

then

Then
$$\{\mathcal{O}_{2}(+-k|4)\}_{k\in\mathbb{Z}} = \{2\mathcal{O}_{0}(4+-k)\}_{k\in\mathbb{Z}}$$
is an $1-basis$ for 134π .

For any
$$j^{20}$$
, if we define $\mathcal{B}_{j}(t) = 2^{j/2} \mathcal{B}_{o}(2^{j}t)$

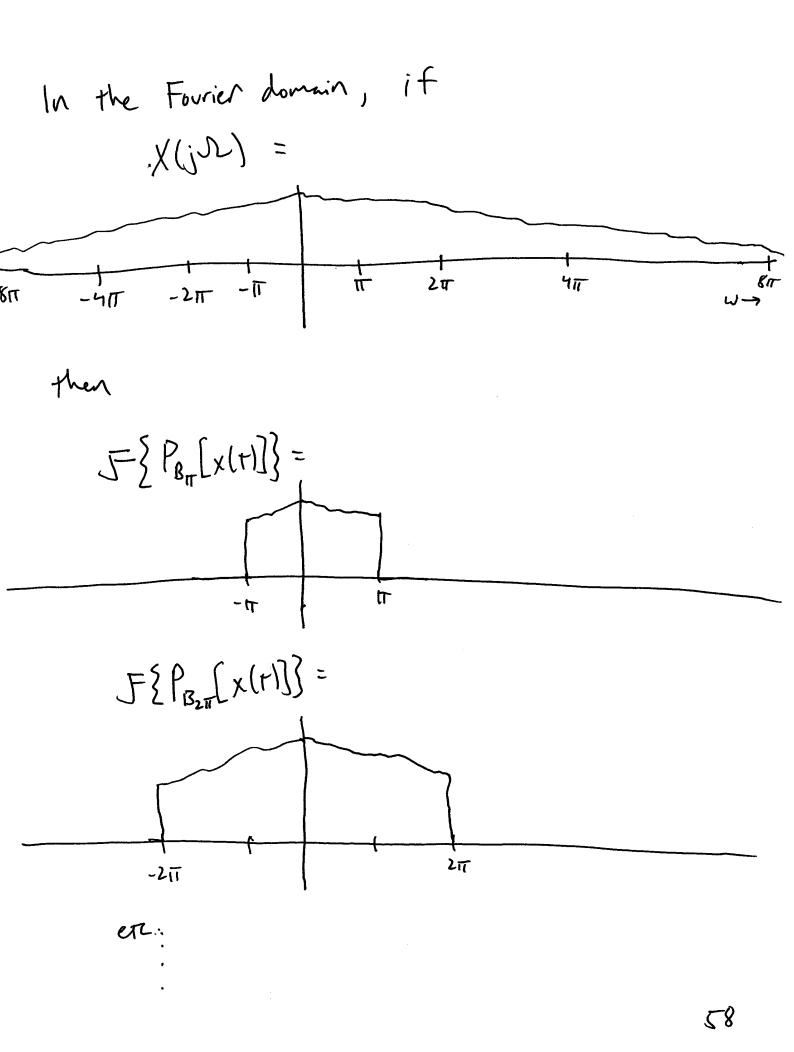
then
$$\{ \mathcal{O}_{j}(t-2^{-j}k) \}_{k \in \mathbb{Z}} = \{ 2^{j/2} \mathcal{O}_{\delta}(2^{j}t-k) \}_{k \in \mathbb{Z}}$$
 is an $1-basis$ for $B_{2^{j}\pi}$.

It should be clear that

$$P_{B_{2}i\pi}\left[\chi(t)\right] = \sum_{k=-\infty}^{\infty} \langle \chi(t), \mathcal{O}_{i}(t-2^{-i}k) \rangle \mathcal{O}_{i}(t-2^{-i}k)$$

gets closer to
$$x(t)$$
 as $j \to \infty$

$$\lim_{j \to \infty} P_{B_{in}}[x(t)] = x(t)$$



Notes by J. Romberg – January 8, 2012 – 16:47

It is clear that BTC BZT. Now consider the difference between the spaces Bot and Bet Wn = BzT O BT = { Signals in BzTT that are orthogonal to BTT } = { Signals whose Fourier transform is supported on IT & |w| \$ 2TT } We can also find an I-basis for Wo. Just as it is true that integer shifts of $\mathcal{O}_{o}(r) = \mathcal{F}^{-1} \left\{ \begin{array}{c} -\pi \\ -\pi \end{array} \right\}$ are an 1-basis for BTT, the integer shifts of $Y_{o}(t) = F^{-1} \left\{ \frac{1}{-2\pi - \pi} \right\}$ are an I-basis for Wo.

59

(You will solidity this on the next HW.)

$$4.(t) = 2 \cos\left(\frac{3\pi t}{2}\right) \frac{\sin(\pi t/2)}{\pi t}$$

Given an arbitrary
$$X(t) \in L_2(IR)$$
, we now have two ways to write the approximation of $X(t)$ in B_{2TT} :

$$P_{\beta_{2\pi}}[\chi(t)] = \sum_{k=-\infty}^{\infty} \langle \chi(t), \vartheta_{1}(t-kl_{2}) \rangle \vartheta_{1}(t-kl_{2})$$

and

PBZT [
$$X(t)$$
] = PBT [$X(t)$] + Pw_o [$X(t)$]

(this equality holds since $B_{TT} \perp W_{o}$,

i.e. everything in B_{TT} is L to everything

in W_{o})

= $\sum_{K=-\infty}^{\infty} \langle X(t), P_{o}(t-K) \rangle P_{o}(t-K) + \sum_{K=-\infty}^{\infty} \langle X, P_{o}(t-K) \rangle P_{o}(t+K)$

Likewise, we can define

$$W_{j} = B_{2^{j+}\pi} \bigoplus B_{2^{j+}\pi}$$

$$= \left\{ \text{ Signals bandlimited to lule } \left[2^{j}\pi, 2^{j+1}\pi \right] \right\}$$

If we set

$$Y_{j}(t) = 2^{j}2 Y_{0}(2^{j}t)$$

then

$$\left\{ Y_{j}(t-2^{-j}k) \right\}_{k\in\mathbb{Z}} = \left\{ 2^{j}2 Y_{0}(2^{j}t-k) \right\}_{k\in\mathbb{Z}}$$
is an $1-basis$ for W_{j} .

Now we are write for any $x(t) \in L_{2}(\mathbb{R})$

$$x(t) = P_{B_{\pi}}[x(t)] + P_{W_{0}}[x(t)] + P_{W_{1}}[x(t)] + P_{W_{2}}[x(t)] + P_{W_{2}}[x(t)]$$

$$P_{B_{\pi}}[x(t)]$$

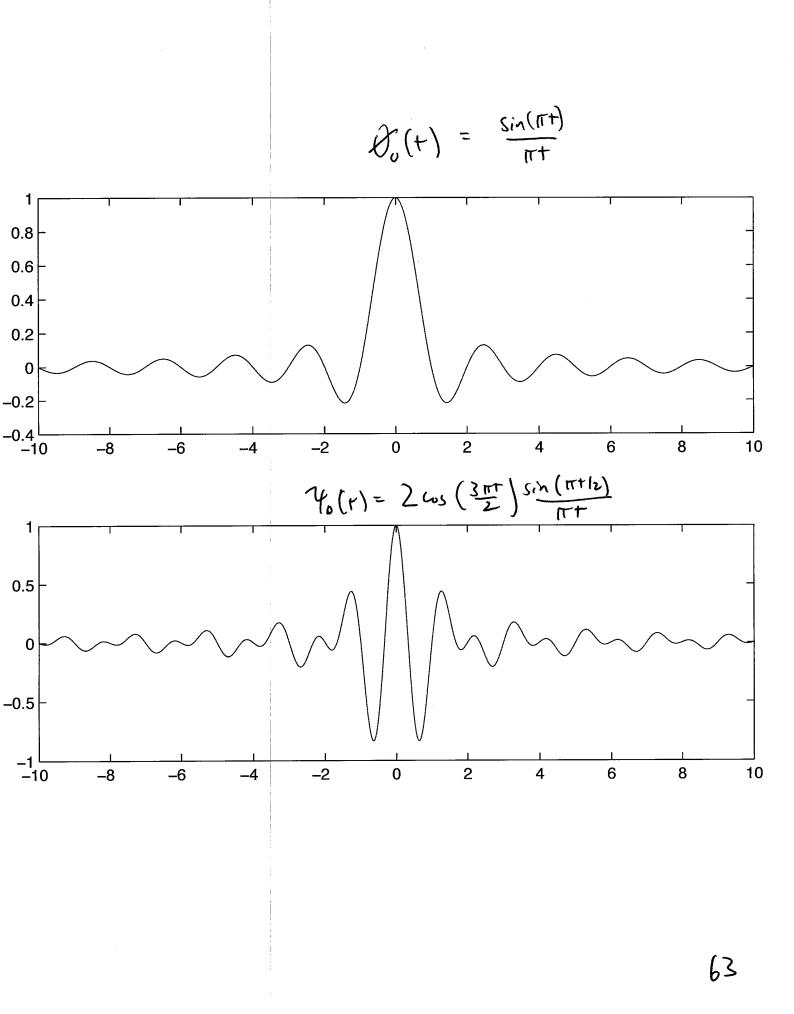
or
$$\chi(t) = \sum_{k=-\infty}^{\infty} \langle \chi(t), \mathcal{O}_{0}(t-k) \rangle \mathcal{O}_{0}(t-k)$$

 $+ \sum_{k=-\infty}^{\infty} \langle \chi(t), \mathcal{V}_{0}(t-k) \rangle \mathcal{V}_{0}(t-k)$
 $+ \sum_{k=-\infty}^{\infty} \langle \chi(t), \mathcal{V}_{1}(t-k|2) \rangle \mathcal{V}_{1}(t-k|2)$
 $+ \sum_{k=-\infty}^{\infty} \langle \chi(t), \mathcal{V}_{2}(t-k|4) \rangle \mathcal{V}_{2}(t-k|4)$

$$= \sum_{k=-\infty}^{\infty} \langle \chi(t), \mathcal{N}_{0,k}(t) \rangle \mathcal{N}_{0,k}(t) + \sum_{j\geq 0}^{\infty} \sum_{k=-\infty}^{\infty} \langle \chi(t), \mathcal{V}_{j,k}(t) \rangle \mathcal{V}_{j,k}(t)$$

where
$$Q_{0,K}(t) = Q_0(t-K)$$

 $Y_{ijK}(t) = Z^{il2}Y_0(Z^{i}t-K)$



Notes by J. Romberg – January 8, 2012 – 16:47

Example: Haar Wavelets

Above, we defined an 1-basis by looking at the difference between approximations at subsequent dyadic scales.

We can do something very similar with piecewise constant functions in the time domain.

Let So = { Signals that are piecewise constant between the integers} 7 e member of So

 $\mathcal{O}_{0}(t) = \begin{cases} | 0 \le t \le 1 \\ 0 \text{ else} \end{cases}$

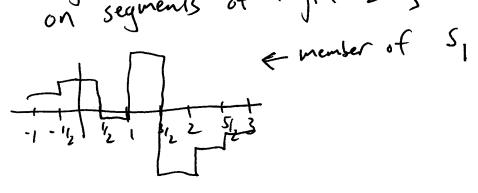
It is clear that { Øv(+-k)} ken is an 1-basis for So

64

Also, given an arbitrary x(+) ELz(IR), We can find the best piecewise constant approximation to x(t) by projecting onto S_0 :

$$P_{s,[\chi(t)]} = \sum_{k=-\infty}^{\infty} \langle \chi(t), \beta_{s}(t-k) \rangle \beta_{s}(t-k)$$

Now let



If we set $\mathcal{O}_{i}(t) = 2^{il_2} \mathcal{O}_{o}(2^{i}t)$

then
$$\{O_j(t-2^{-j}k)\}_{k\in\mathbb{Z}} = \{2^{-j}l_2O_o(2^{j}t-k)\}_{k\in\mathbb{Z}}$$
is an $1-basis$ for S_j .

lim $P_{S_i}[x(t)] = x(t)$ $j = \infty$ Si [x(t)] = x(t)From Burrus et al: Also

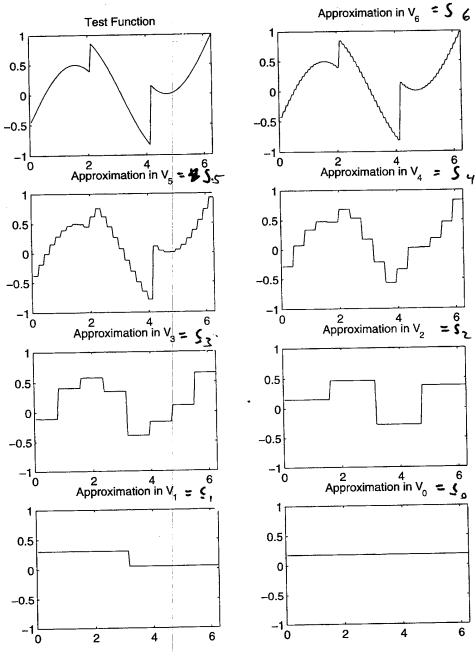


Figure 2.15. Haar Function Approximation in \mathcal{V}_j

Now Set

$$W_0 = S_1 \oplus S_0$$

= {signals in S, that are \bot to S_0 }
and more generally
 $W_j = S_{j+1} \oplus S_j$

Wo contains signals that are piecewise constant between the half integers and have zero mean on intervals between the integers

member of who will be the constant of the integers are the integers of the constant of the constant are piecewise constant to the constant are piecewise constant.

If we define $Y_0(t) = \begin{cases} 1 & 0 \le t \le 1/2 \\ -1 & 1/2 \le t \le 1 \end{cases}$ else

Yo(t)

then

{4 (+-k)} KEZ is an I-basis for Wo

If we define
$$Y_{j}(t) = \sum_{j=1}^{j} Y_{0}(2^{j}t)$$

then
$$\{Y_{j}(t-2^{-j}k)\}_{k\in\mathbb{Z}} = \{2^{j}2^{j}2^{j}(2^{j}t-k)\}_{k\in\mathbb{Z}}$$
 is an $1-basis$ for W_{j} .

We can now write any X(t) ELZ(IR) as

$$\chi(t) = P_{s_0}[\chi(t)] + P_{w_0}[\chi(t)] + P_{w_1}[\chi(t)] + P_{w_2}[\chi(t)] + P_{w_2}[\chi(t)]$$

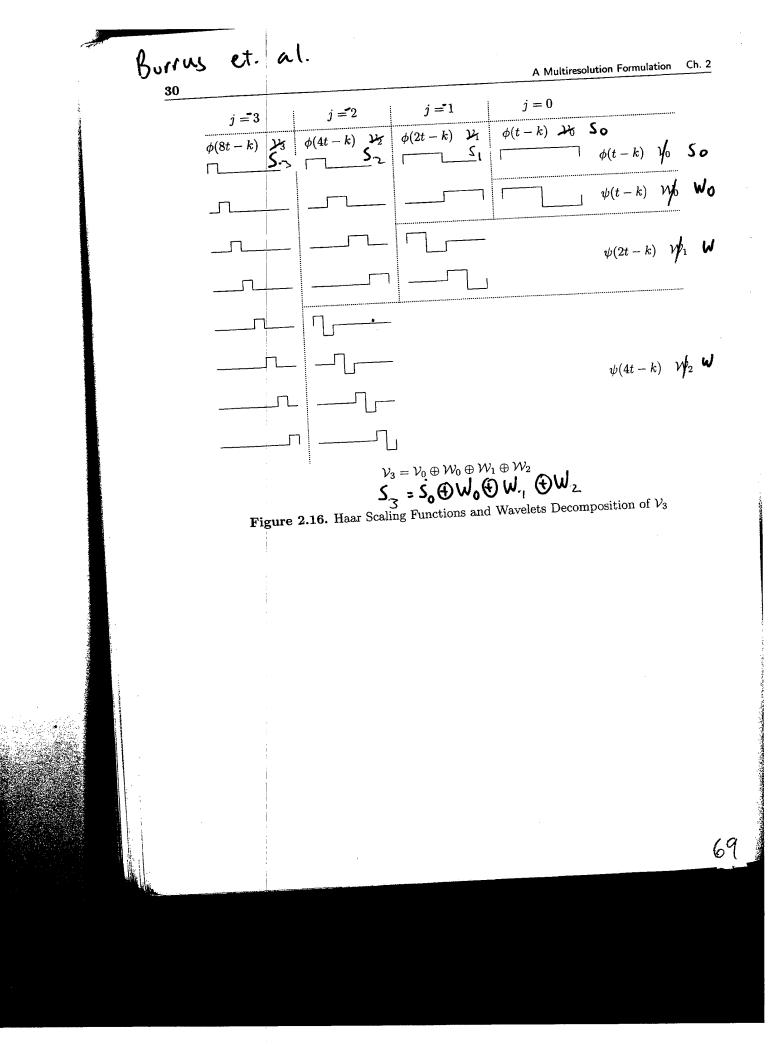
$$P_{s_1}[\chi(t)]$$

$$P_{s_2}[\chi(t)]$$

i.e.
$$\chi(+) = \sum_{k=-\infty}^{\infty} \langle \chi(+), \mathcal{B}_{0,k}(+) \rangle \mathcal{B}_{0,k}(+) + \sum_{j\geq 0}^{\infty} \langle \chi, Y_{j,k} \rangle Y_{j,k}(+)$$

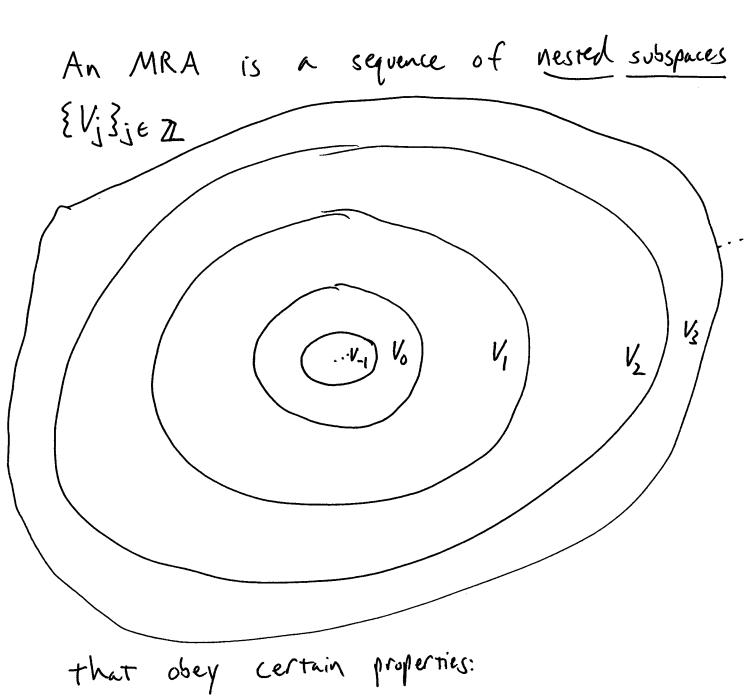
where as before

68



Multiresolution Approximations (MRAS)

Abstracting the key properties of the Shannon and Haar wavelet systems above will give us a general framework for orthonormal wavelet transforms.



- () V_j C V_{j+1}
- $\lim_{j\to\infty} V_j = L_2(\mathbb{R})$ $j\to\infty \qquad j=\infty$ (i.e. $\lim_{j\to\infty} P_{V_j}[x(t)] = x(t) \quad \forall \ x(t) \in L_2(\mathbb{R})$)
- (3) $\lim_{j\to-\infty} V_j = \{0\}$ (only the zero function)
- (6) There exists a $\mathcal{O}_{0}(t)$ \in V_{0} such that $\{\mathcal{O}_{0}(t-k)\}_{k\in\mathbb{Z}}$ is an 1-busis for V_{0}

Consined with the other conditions, 6 tells us that {2ih o(2i+-K)} kez is an I-busis for Vi As such, we can compute the best approximation to X(+) in Vi wing $P_{V_{i}}\left[\chi(t)\right] = \sum_{k=-\infty}^{\infty} \langle \chi(t), \theta_{j,k}(t) \rangle \theta_{j,k}(t)$

where

Moreover, if we define

$$W_j = V_{j+1} \bigoplus V_j$$

= everything in V_{j+1} that is \bot to V_j

it can be shown that there exists a $Y_0(+) \in W_0$

Combined with the properties above, this means {2^{i/2} 4, (2ⁱ+-k)} ** EZ is an 1-basis As before, we will use the notation 4, (+) = 2 1/2 4 (2 1+ - K).

Nomenclature

The V; are called scaling spaces, and the Oj, k(+) are called scaling functions.

The W; are called wavelet spaces (or detail spaces) and the 4j, k(+) are called wavelets.

Decomposing L2(IR)

We also have that

 $L_{2}(IR) = \bigvee_{0} \bigoplus W_{0} \bigoplus W_{1} \bigoplus W_{2} \bigoplus W_{3} \bigoplus \cdots$

or in general we can start at any scale J. $L_2(IR) = V_J \oplus W_J \oplus W_{J+1} \oplus W_{J+2} \oplus \cdots$

We can decompose any $x(t) \in L_2(IR)$ as $x(t) = P_{V_0}[x(t)] + P_{W_0}[x(t)] + P_{W_1}[x(t)] + P_{W_2}[x(t)] + \cdots$ $= \sum_{K=-\infty}^{\infty} \langle x(t), P_{O_1K}(t) \rangle P_{O_1K}(t) + \sum_{j \geq 0}^{\infty} \langle x(t), V_{j,j,K}(t) \rangle V_{j,k}(t)$ $= \sum_{K} C_{O,K} P_{O,K}(t) + \sum_{j \geq 0}^{\infty} \sum_{K=-\infty}^{\infty} d_{j,K} V_{j,k}(t)$

We call

Cj.k = < X, Bj.k > the scaling coefficients
at scale j

and dik = <x, 4j, k > the wavelet welficients at scale j

The mapping

is called the wavelet transform.

Again, we can initiate this at any scale J, Not just J=0.

Moving from scale to scale

Let's look closely at the relationship between scales j=0 and j=-1 (what we learn can scales j=0 and j=-1 for any j). be immediately generalized to j and j-1 for any j).

First, since 1, cVo, we know we an write D_(+) = \frac{1}{2} \(\text{\(\frac{1}{2}\)} \)

as a linear combination of { Bo(+-R) 3R.

That is, a dilated version of \$0(t) Can be written as a combination of shifts of 80(+)

$$\frac{\partial_{-1}(+)}{\partial_{2}} = \frac{1}{\sqrt{2}} \mathcal{O}_{0}(+12) = \frac{1}{\sqrt{2}} \mathcal{O}_{0}(+12), \mathcal{O}_{0}(+-2) = \frac{1}{\sqrt{2}} \mathcal{O}_{0}(+12), \mathcal{O}_{0}(+-2) = \frac{1}{\sqrt{2}} \mathcal{O}_{0}(+12) = \frac{1}{\sqrt{2}} \mathcal{O}_{0}(+$$

The sequence h[l] contains the expansion wefficients for Q_(t) in Vo.

It is true (and you can check this) that the same h[l] can be used at every scale. In general

$$\mathcal{S}_{j-1,0}(t) = \frac{1}{\sqrt{2}} \mathcal{S}_{j,0}(t) = \sum_{\ell} h[\ell] \mathcal{S}_{j,\ell}(t)$$

We can do the same thing for any other shift of $O_{1}(t)$

$$\mathcal{Q}_{-1,K}(+) = \frac{1}{\sqrt{2}} \mathcal{Q}_{0}(+|2-k|)$$

$$= \underbrace{\sum_{\ell} \left\langle \frac{1}{\sqrt{2}} \mathcal{Q}_{0}(+|2-k|), \mathcal{Q}_{0}(+-\ell) \right\rangle}_{\mathcal{Q}_{0}(+-\ell)} \mathcal{Q}_{0}(+-\ell)$$

$$= \underbrace{\sum_{\ell} \left\langle \frac{1}{\sqrt{2}} \mathcal{Q}_{0}(+|2|), \mathcal{Q}_{0}(+-\ell+2k) \right\rangle}_{\mathcal{Q}_{0}(+-\ell)} \mathcal{Q}_{0}(+-\ell)$$

$$= \underbrace{\sum_{\ell} \left\langle \frac{1}{\sqrt{2}} \mathcal{Q}_{0}(+|2|), \mathcal{Q}_{0}(+-\ell) \right\rangle}_{\mathcal{Q}_{0}(+-\ell)}$$

$$= \underbrace{\sum_{\ell} \left\langle \frac{1}{\sqrt{2}} \mathcal{Q}_{0}(+|2|), \mathcal{Q}_{0}(+-\ell) \right\rangle}_{\mathcal{Q}_{0}(+-\ell)}$$

and in general

$$\mathcal{O}_{j-1,k}(t) = \sum_{\ell} h[\ell-2k] \mathcal{O}_{j,\ell}(t)$$

These relationships give us a very vice way to move from an approximation in V_0 $\left[V_0\left[X(t)\right] = \sum_{K} C_{0,K} \mathcal{S}_{0,K}(t)\right]$ to an approximation in V_1 $\left[V_1\left[X(t)\right] = \sum_{K} C_{1,K} \mathcal{K}_{1,K}(t)\right]$

Notice:

$$C_{-1,K} = \langle x(t), \theta_{-1,K}(t) \rangle$$

$$= \langle x(t), \xi_{0}h(l-2k)\theta_{0,e}(t) \rangle$$

$$= \xi_{0}h(l-2k)\langle x(t), \theta_{0,e}(t) \rangle$$

$$= \xi_{0}h(l-2k)\langle x(t), \theta_{0,e}(t) \rangle$$

$$= \xi_{0}h(l-2k)\langle x(t), \theta_{0,e}(t) \rangle$$

To make this look a little more familiar, we write

$$C_{-1}[k] = \sum_{k} h(k-2k) c_{0}[k]$$

$$= \left(C_{0}[k] + h[k]\right) down sampled by 2$$

where Ho has impulse response

$$h_{o}[k] = h[-k] = \langle Q_{1}(t), Q_{0}(t+k) \rangle$$

$$= \langle J_{0}(t), Q_{0}(t+k) \rangle$$

Similarly, since 4-1(+) EVO

$$\Psi_{-1}(t) = \sum_{\ell} \{ \Psi_{-1}(t), \emptyset_{0}(t-\ell) \} \emptyset(t-\ell)$$

and
$$Y_{-1,k}(t) = \sum_{\ell} \langle \frac{1}{\sqrt{2}} Y_{0}(t_{2}-k), \vartheta_{0}(t-\ell) \rangle \vartheta_{0}(t-\ell)$$

$$= \sum_{\ell} g[\ell-2k] \vartheta_{0}(t-\ell)$$

$$P_{W_{-1}}[x(t)] = \sum_{K} d_{-1,K} \Psi_{-1,k}(t)$$

where

$$d_{-1,k} = \langle x(t), Y_{-1,k}(t) \rangle$$

$$= \langle x(t), \sum_{\ell} g(\ell-2k) \mathcal{O}_{0}(t-\ell) \rangle$$

$$= \sum_{\ell} g[\ell-2k] \langle x(t), \mathcal{O}_{0}(t-\ell) \rangle$$

$$= \sum_{\ell} g[\ell-2k] C_{0}[\ell]$$

50

$$C_0[k] \longrightarrow [H_1] \longrightarrow d_1[k]$$

where H, has impulse response $h_{i}[k] = g[-k] = \left\langle \frac{1}{\sqrt{2}} Y_{i}(t_{2}), \theta_{o}(t+k) \right\rangle$

Thus we can interpret

$$P_{V_0}[x(t)] = P_{V_1}[x(t)] + P_{W_1}[x(t)]$$
as a filter bank
$$H_0 \longrightarrow U_2 \longrightarrow C_1[k]$$

$$C_0[k] \longrightarrow H_1 \longrightarrow U_2 \longrightarrow C_1[k]$$
and more generally
$$P_{V_1}[x(t)] = P_{V_1}[x(t)] + P_{W_1-1}[x(t)]$$
as
$$C_1[k] \longrightarrow H_0 \longrightarrow U_2 \longrightarrow G_1[k]$$

$$C_1[k] \longrightarrow H_1 \longrightarrow U_2 \longrightarrow G_1[k]$$

Wavelet basis functions and filter banks

(iven \$5.(+), 4.(+) we know how to construct the corresponding filters:

$$h[n] = \langle \varnothing_{-1}(t), \varnothing_{0}(t-n) \rangle$$

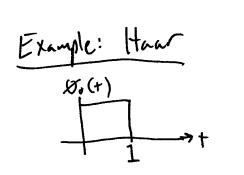
$$= \langle \frac{1}{\sqrt{2}}, \varnothing_{0}(t|u), \varnothing_{0}(t-n) \rangle$$

$$= \langle \Psi_{1}(t), \varnothing_{0}(t-n) \rangle$$

$$= \langle \frac{1}{\sqrt{2}}, \Psi_{1}(t|u), \varnothing_{0}(t-n) \rangle$$

$$= \langle \frac{1}{\sqrt{2}}, \Psi_{1}(t|u), \varnothing_{0}(t-n) \rangle$$

and set



$$\frac{\partial_{-1}(+)}{\partial z} = \frac{1}{\sqrt{2}} \frac{\partial_{0}(+)}{\partial z} + \frac{1}{\sqrt{2}} \frac{\partial_{0}(+-1)}{\partial z} + \frac{\partial_{$$

$$Y_{-1}(t) = \frac{1}{\sqrt{2}} \mathcal{O}_{0}(t) - \frac{1}{\sqrt{2}} \mathcal{O}_{1}(t)$$

$$h_0(n) = \int_0^{n}$$
 $h_1(n) = \int_0^{n}$

This is just a shifted version of what we used for the Haar filkerbank earlier in the course (h, is also negated, but this makes no effective difference since its symmetric).

Example: Shannon (see the homework)

This connection between basis functions and filter banks also goes the other way. Given a CMF, there is a corresponding scaling function $O_o(t)$ and wavelet $V_o(t)$.

In general, given ho[n] (and the corresponding h_[n]), Bo(t) (and Yo(t)) cannot be written down explicitly—rather it is defined by the recursion

$$\frac{1}{\sqrt{2}} \mathcal{O}_{0} \left(+ 1_{2} \right) = \sum_{e} h(\underline{l}) \mathcal{O}_{0} \left(+ e \right)$$

$$= h_{0}[-e]$$

In the Fourier domain, this gives us

$$\nabla \overline{\Sigma}_{o}(j2\pi) = \Sigma_{o}h[\ell] e^{-j\pi\ell} \overline{\Sigma}_{o}(j\pi)$$

$$= \left(\overline{\mathbf{F}}_{o}(\mathbf{j} \mathbf{N}) - \left(\mathbf{F}_{e} \mathbf{h}(\mathbf{l}) e^{-\mathbf{j} \mathbf{n} \mathbf{l}} \right) \right)$$

i.e.

$$\overline{\Phi}_{o}(jR) = \frac{1}{\sqrt{n}} H(e^{jM_{2}}) \, \overline{\Phi}_{o}(jM_{2})$$

$$= \frac{1}{\sqrt{n}} H(e^{jM_{2}}) \frac{1}{\sqrt{n}} H(e^{jM_{1}}) \, \overline{\Phi}_{o}(jM_{1})$$

$$= \frac{1}{\sqrt{n}} H(e^{jM_{2}}) \frac{1}{\sqrt{n}} H(e^{jM_{1}}) \cdot \frac{1}{\sqrt{n}} H(e^{jM_{1}}) \, \overline{\Phi}_{o}(jM_{1})$$

$$= \left(\frac{P}{P_{-1}} \frac{1}{\sqrt{n}} H(e^{j2^{-P_{1}}})\right) \cdot \overline{\Phi}_{o}(j2^{-P_{1}})$$

$$= \left(\frac{P}{P_{-1}} \frac{1}{\sqrt{n}} H(e^$$

So given ho[n] (and hence h[n] and hence H(eir)) we can define of, (+) in the Forier domain using this infinite product.

Given
$$\mathcal{O}_{0}(t)$$
, we can always construct

$$\Psi_{0}(t) = \sum_{\ell} g[\ell] \mathcal{O}_{1}(t-\ell)$$

$$= \sqrt{2} \sum_{\ell} g[\ell] \mathcal{O}_{5}(2t-\ell)$$
where

g[e]=h,[-e]= (-1) -e ho[1-e].

				-			-
		n	$h_p[n]$, [$h_p[n]$	
Δ.,		0	482962913145	i		**p[(*)	
04	p=2	i	836516303738	p=8	0	.054415842243	
1.		2	.224143868042	h=0	ĭ	.312871590914	
4		3	129409522551	1	2	.675630736297	
		 -		i	3	.585354683654	
	p=3	0	.332670552950		4	015829105256	
D6	. 1	1	.806891509311	į	5	284015542962	
,	1 1	2	.459877502118	1	6	.000472484574	
احا	1	3	135011020010		7	.128747426620	
		4	085441273882		8	017369301002	_
		5	.035226291882		9	04408825393	FA
•		0	.230377813309		10	.013981027917	1.
•	p≈4	ĭ	714846570553		11	004870352993	MA
	1	2	,630880767930		12 13	000391740373	٠.
0 17	1	3	027983769417		14	.000675449406	
ett.	1	4	187034811719		15	000117476784	CW
	ļ '	5	.030841381836				
	Į.	6	.032883011667	p=9	lo	.038077947364	
	1	7	010597401785	, P	i	.243834674613	
		1		•	2	.604823123690	
	p = 5	0	.160102397974	ì	3	.657288078051	
	1	1	.603829269797		4	.133197385825	
	1	2	.724308528438		5	293273783 279	
	1	3	.138428145901		6	096840783223	
	ì	4	242294887066 032244869585	1	7	.148540749338	
,	ļ	5	032244869363	ŀ	8	.030725681479	
	į	6	006241490213		9	067632829061 .000250947115	
	1	8	012580751999	il	10	.002361662124	
	1	وا	.003335725285	1	11 12	004723204758	
	<u></u>	 		 	13	004281503682	
	p=6	lo	.111540743350	ll .	14	.001847646883	
	1 "	1	.494623890398	1	15	.000230385764	ĺ
		2	.751133908021	11	16	000251963189	
	1	3	,315250351709	li	17	.000039347320	'
	1	4	226264693965	II	+-	***************************************	ŀ
	1	5	129766867567 .097501605587	p = 10	0	.026670057901	1
	1	6	.027522865530	H	1	.188176800078	1
	1	8	031582039317	ll .	3	.527201188932 .688459039454	1
	1	١٥	.000553842201	1	1 4	281172343661	i
		10	.004777257511	1	5	249846424327	ļ.
	1	lii	001077301085	li	6	195946274377	1
				1	1 7	.127369340336	1
	p=7	0	.077852054085	li .	8	.093057364604	Ì
	1	1	396539319482	1 1	9	071394147166	ł
	1	2	.729132090846 .469782287405		10	029457536822	1
	[3	- 143906003929	H	11	.033212674059 .003606553567	1
	-	5	224036184994	ii .	12		1
		6	.071309219267	11	13		1
	1	1 7	,080612609151	1	15		1
	-	8	038029936935	ii.	16	0000000000000000	1
	1	9	016574541631	II.	17		1
	1	10	.012550998556	Ĥ	18	.000093588670	1
	1	11	.000429577973	1	19		1
	1	12	001801640704				

These are the Daubechies CMF filter wells from Sec I of the notes. Each corresponds to a different MRA/wavelet system,

called ~ D-2p"

DZ= Huar

iscrete Moments ī

tems in general as 1) and (5.10) that $0) = m_1(0) = 0$ of : Orthonormality equires m(0) = 1legrees of freedom 5th-6 we have two s we have exactly N/2 zero wavelet ationships among xplained in (6.52)

s of the wavelets ions. Figures 6.1 , 10, 12, 16, 20, 40. Chapter 6. The id in [Dau92] or For N=2, the ble; for N = 6, it for longer h(n).

oments that are e the degrees of

from Burns et al.) 81 Sec. 6.3. Daubechies' Method for Zero Wavelet Moment Design

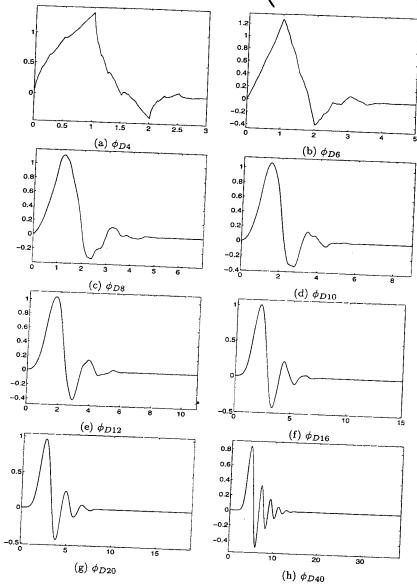


Figure 6.1. Daubechies Scaling Functions, $N=4,6,8,\ldots,40$

freedom to maximize the differentiability of $\varphi(t)$ rather than maximize the zero moments. This is not easily parameterized, and it gives only slightly greater smoothness than the Daubechies

86

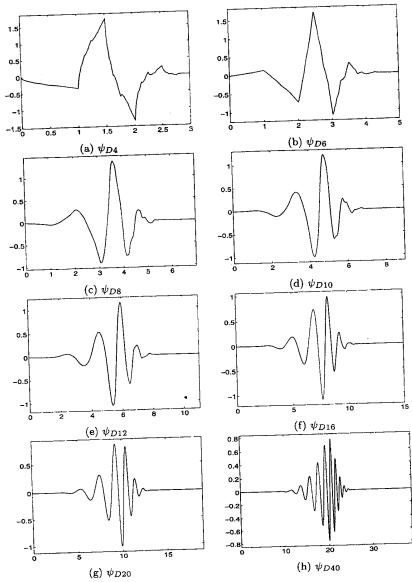


Figure 6.2. Daubechies Wavelets, $N = 4, 6, 8, \dots, 40$

Examples of Daubechies scaling functions resulting from choosing different factors in the spectral factorization of $|H(\omega)|^2$ in (6.18) can be found in [Dau92].

2.7 Examples of Wavelet Expansions

In this section, we will try to show the way a wavelet expansion decomposes a signal and what the components look like at different scales. These expansions use what is called a length-8 Daubechies basic wavelet (developed in Chapter 6), but that is not the main point here. The local nature of the wavelet decomposition is the topic of this section.

These examples are rather standard ones, some taken from David Donoho's papers and web page. The first is a decomposition of a piecewise linear function to show how edges and constants are handled. A characteristic of Daubechies systems is that low order polynomials are completely contained in the scaling function spaces V_j and need no wavelets. This means that when a section of a signal is a section of a polynomial (such as a straight line), there are no wavelet expansion coefficients $d_j(k)$, but when the calculation of the expansion coefficients overlaps an edge, there is a wavelet component. This is illustrated well in Figure 2.6 where the high resolution scales gives a very accurate location of the edges and this spreads out over k at the lower scales. This gives a hint of how the DWT could be used for edge detection and how the large number of small or zero expansion coefficients could be used for compression.

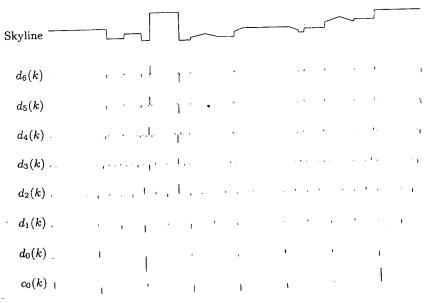


Figure 2.5. Discrete Wavelet Transform of the Houston Skyline, using $\psi_{D8'}$ with a Gain of $\sqrt{2}$ for Each Higher Scale

Figure 2.6 shows the approximations of the skyline signal in the various scaling function spaces V_j . This illustrates just how the approximations progress, giving more and more resolution at higher scales. The fact that the higher scales give more detail is similar to Fourier methods, but the localization is new. Figure 2.7 illustrates the individual wavelet decomposition by showing

signal and what called a length-8 point here. The

s papers and web ges and constants als are completely lat when a section wavelet expansion lps an edge, there resolution scales lower scales. This e number of small

with a Gain of $\sqrt{2}$

ing function spaces more resolution at urier methods, but osition by showing the components of the signal that exist in the wavelet spaces W_j at different scales j. This shows the same expansion as Figure 2.6, but with the wavelet components given separately rather than being cumulatively added to the scaling function. Notice how the large objects show up at the lower resolution. Groups of buildings and individual buildings are resolved according to their width. The edges, however, are located at the higher resolutions and are located very accurately.

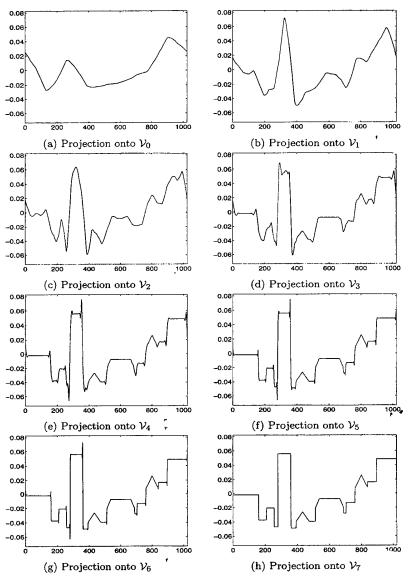


Figure 2.6. Projection of the Houston Skyline Signal onto $\mathcal V$ Spaces using $\phi_{D8'}$

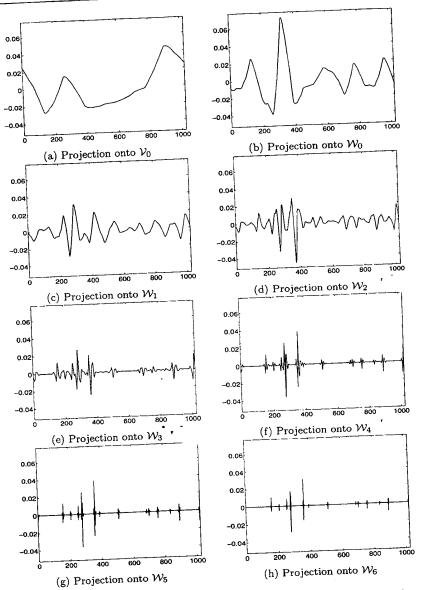
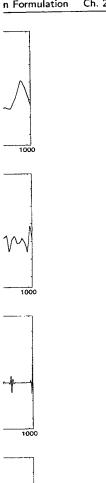


Figure 2.7. Projection of the Houston Skyline Signal onto $\mathcal W$ Spaces using $\psi_{DS'}$

The second example uses a chirp or doppler signal to illustrate how a time-varying frequency is described by the scale decomposition. Figure 2.8 gives the coefficients of the DWT directly as a function of j and k. Notice how the location in k tracks the frequencies in the signal in a way the Fourier transform cannot. Figures 2.9 and 2.10 show the scaling function approximations and the wavelet decomposition of this chirp signal. Again, notice in this type of display how the "location" of the frequencies are shown.



s using $\psi_{D8'}$

e-varying frequency he DWT directly as the signal in a way ion approximations e of display how the

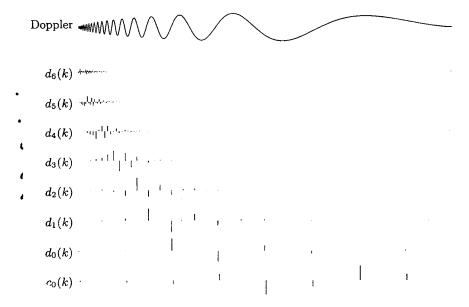


Figure 2.8. Discrete Wavelet Transform of a Doppler, using $\psi_{D8'}$ with a gain of $\sqrt{2}$ for each higher scale.

2.8 An Example of the Haar Wavelet System

In this section, we can illustrate our mathematical discussion with a more complete example. In 1910, Haar [Haa10] showed that certain square wave functions could be translated and scaled to create a basis set that spans L^2 . This is illustrated in Figure 2.11. Years later, it was seen that Haar's system is a particular wavelet system.

If we choose our scaling function to have compact support over $0 \le t \le 1$, then a solution to (2.13) is a scaling function that is a simple rectangle function

$$\varphi(t) = \begin{cases} 1 & \text{if } 0 < t < 1\\ 0 & \text{otherwise} \end{cases}$$
 (2.42)

with only two nonzero coefficients $h(0) = h(1) = 1/\sqrt{2}$ and (2.24) and (2.25) require the wavelet

$$\psi(t) = \begin{cases} 1 & \text{for } 0 < t < 0.5 \\ -1 & \text{for } 0.5 < t < 1 \\ 0 & \text{otherwise} \end{cases}$$
 (2.43)

with only two nonzero coefficients $h_1(0) = 1/\sqrt{2}$ and $h_1(1) = -1/\sqrt{2}$.

 \mathcal{V}_0 is the space spanned by $\varphi(t-k)$ which is the space of piecewise constant functions over integers, a rather limited space, but nontrivial. The next higher resolution space \mathcal{V}_1 is spanned by $\varphi(2t-k)$ which allows a somewhat more interesting class of signals which does include \mathcal{V}_0 . As we consider higher values of scale j, the space V_j spanned by $\varphi(2^jt-k)$ becomes better able to approximate arbitrary functions or signals by finer and finer piecewise constant functions.

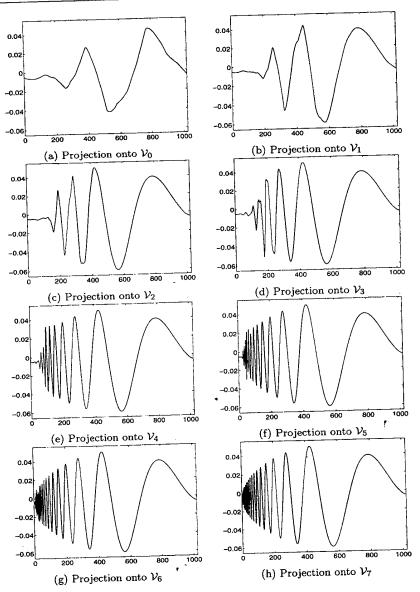


Figure 2.9. Projection of the Doppler Signal onto $\mathcal V$ Spaces using $\phi_{D8'}$

92

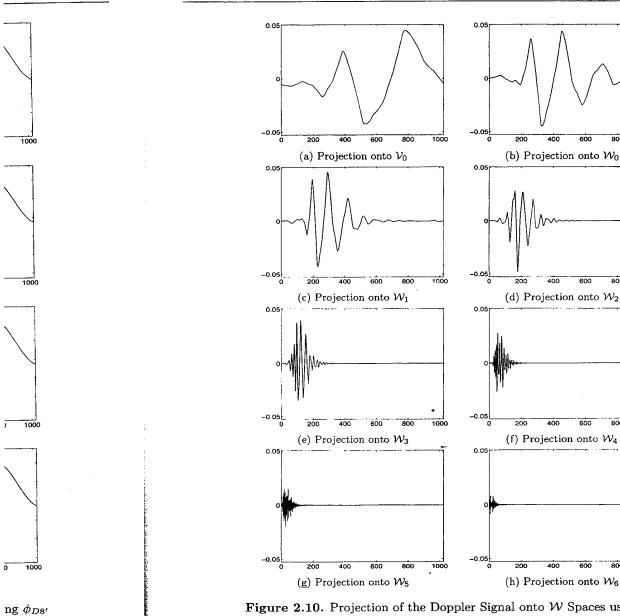


Figure 2.10. Projection of the Doppler Signal onto W Spaces using $\psi_{D8'}$

800

1000

The Daubechies wavelets, illustrated on the past few pages, have the following properties

() { Do (t-k) } KER U { 2ⁱ¹² 40 (2ⁱt-k) } j20, KER is an <u>1-basis</u> for L₂(IR).

This basically follows from the "perfect reconstruction" and "orthogonality" conditions for the corresponding CMF.

(2) Vanishing moments, 4 (t) in the D2p

System has p vanishing moments. This

System has p vanishing moments. This

for q=0,-,p-1 components in the wavelet spaces Wj.

It follows from hi[n] having vanishing moments

in discrete time (see homework)

3 Support size. The scaling function Bo(t) and wavelet function are local in time. Their support size is 2p-1 (closely related to the lengths of the FIR filters ho, h,).

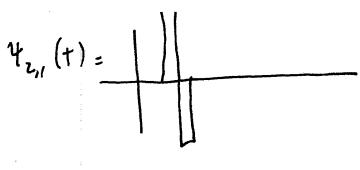
What are wavelets good for?

At the end of the day, the wavelet transform is just another way to "take apart" a Signal.

For Xc(t) on [0,1]

$$X_{c}(t) = C_{\bullet,\bullet}(t) + \sum_{j\geq 0} \sum_{K} d_{j}[K] \Psi_{j,K}(t)$$

$$\theta_{o,o}(r) = \frac{1}{1}$$



To Fourier Series busis functions $X_{c}(t) = \sum_{k} a_{k} e^{j2\pi kt}$ $X_{c}(t)$

$$X_c(t) = \sum_{k} a_k e^{j2\pi k t}$$

(real paris)

Why would we want to use wavelets instead of sinusoids (which we already know a lot about)?

$$S_{ay}$$
 $X_{c}(t) = \frac{1}{\sqrt{3}} \frac{1}{\sqrt{3}}$

Questions:

(Haar)

The wavelet coefficient at scalej and shift k

is dj[k] = (xc, 4j,k)

There are Zi wavelets (Litterent with = 2-i shifts) at scale j.

For fixed j, how many of the dilk? are non-zero?

2) Say Xc(+) =

What is an upper bound on the number of non-zero dj[k]? (Again for a fixed j)

Moral: For piecewise constant functions, very few of the terms in $X_{c}(t) = C_{0,0} S_{0,0}(t) + \sum_{j \geq 0} \sum_{k} d_{j}[k] Y_{j,k}(t)$ are non-zero.

What about the Fourier Series expansion? $X_{c}(t) = \sum_{k} a_{k} e^{j2\pi kt}$

How many of the ax are non-zero?

For piecewise constant signals, Haar approximations, are much more accurate than Fourier approximations.

The energy "compacts" onto a much smaller set in the Haar domain than in the Fourier domain, making piecewise constant functions very "simple".

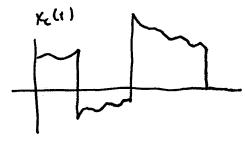
-> We can ignore most of the wavelet crefticients and not lose anything.

This makes wavelets very useful for things like Compression, noise removal, and feature detection. ("nonlinear filtering")

As a general rule, signals which are smooth T

have roughly the same amount of compaction in the wavelet and Fourier domains.

But signals which are piecewise smooth

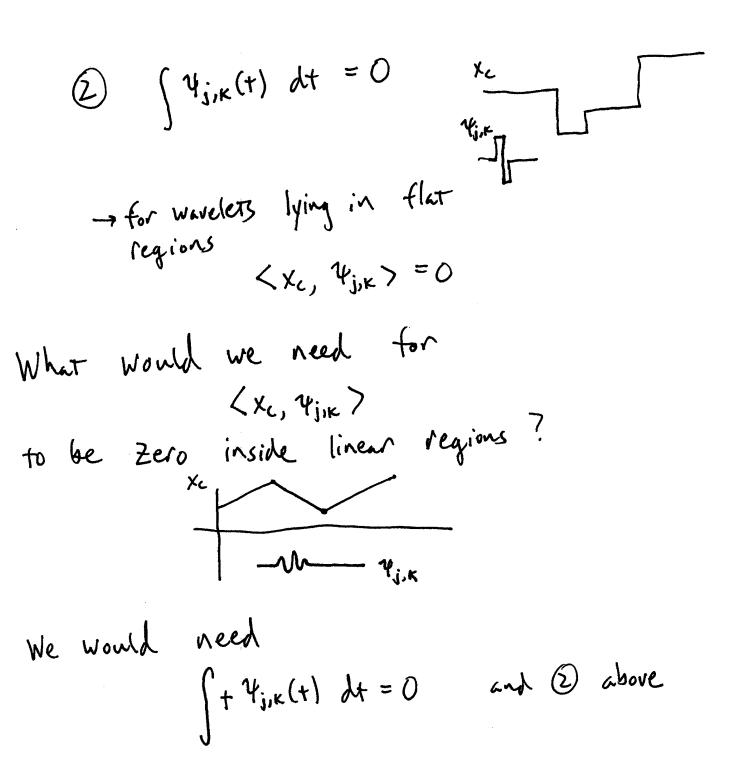


are far more compacted in the wavelet

Piecewise smooth signals come up in all kinds of applications:

- digital photography
- Seismic exploration (i.e. finding oil)
- medical imaging
- radar imaging and remote sensing

Beyond piecewise constant
Hear wavelets: + III - good for piecewise constant - good for piecewise constant
- good for processe linear - not quite as good for piecewise linear
or piecewise polynomial
Hour works well for piecewise constant functions
since 1 it has small support If If If
-> only "touches" edges in a few locations



Xc(t) is polynomial over the support

of degree

4: k(t). Say 4j, K (+).

1+

m=0,...,p \frac{1}{2} + m 4j, k (+) dt = 0 (*)

then it is easy to see that $\langle \chi_c, \Psi_{j,k} \rangle = 0$

If (*) is true, we say V_{jik} has M vanishing moments.

This is a desirable property, as it allows us to extend our approximation results W/ Harr and piecewise constants mts order piecewise Polynomials.

20 Wavelets

Many times we are interested in processing signals that are functions of two spatial variables ("images")

X(s,t) =

above the plane

We can easily move from 10 to 20 $(\chi(t) \in L_2(\mathbb{R})$ To $\chi(s,t) \in L_2(\mathbb{R}^2)$

Winy the notion of <u>separability</u>.

Det: We call a function f(s,t) separable

 $f(s,t) = g(s) \cdot h(t)$

for some g(·) and h(·).

Ex: The 20 Fourier basis functions $e^{j(w_1s+w_2t)}$

are separable. g(s) = h(+) =

Exercise: Suppose
$$g$$
 and h are 10 functions with $g \perp h$ $(\langle g,h\rangle = 0, i.e.)$

Show that the four functions
$$f_1(s,t) = g(s)g(t)$$

$$f_2(s,t) = g(s)h(t)$$

$$f_3(s,t) = h(s)g(t)$$

$$f_4(s,t) = h(s)h(t)$$
are orthogonal.

Let
$$\theta(t) = \frac{1}{1}$$

$$\psi(t) = \frac{1}{1}$$

(Haar Scaling function and wavelet)

Skerch

$$\mathcal{O}^{LL}(s,t) = \mathcal{O}(s) \cdot \mathcal{O}(t)$$

$$\mathcal{V}^{LH}(s,t) = \mathcal{O}(s) \mathcal{V}(t)$$

$$\mathcal{V}^{HL}(s,t) = \mathcal{V}(s) \cdot \mathcal{O}(t)$$

$$\mathcal{V}^{HH}(s,t) = \mathcal{V}(s) \cdot \mathcal{V}(t)$$

on [0,1]2.

Given a 10 wavelet basis for L2(IR):

{00, K} KEZ U {4j, K} j20, KEZ

(or more generally

{OJ,K}KEZU{Yj,K}jzJ, KEZ)

where

$$\mathscr{O}_{j,K}(t) = 2^{j/2} \mathscr{O}(2^{j}t - K)$$

We can create a 2D separable wavelet basis for L2(IR2) with

{\(\mathcal{D}_{0,K_1,K_2}\)}_{\ell_{1,K_1,K_2}\}_{\ell_{1,K_1,K_2}\}_{j \geq 0,K_1,K_2}\}_{j \geq 0,K_1,K_2}\}_{j

Where

$$\theta_{j,\kappa_{1},\kappa_{2}}^{LL}(s,t) = 2^{j}\theta(2^{j}s-k_{1})\theta(2^{j}t-k_{2})$$
 $\Psi_{j,\kappa_{1},\kappa_{2}}^{LH}(s,t) = 2^{j}\theta(2^{j}s-k_{1})\Psi(2^{j}t-k_{2})$
 $\Psi_{j,\kappa_{1},\kappa_{2}}^{HL}(s,t) = 2^{j}\Psi(2^{j}s-k_{1})\theta(2^{j}t-k_{2})$
 $\Psi_{j,\kappa_{1},\kappa_{2}}^{HL}(s,t) = 2^{j}\Psi(2^{j}s-k_{1})\theta(2^{j}t-k_{2})$
 $\Psi_{j,\kappa_{1},\kappa_{2}}^{HH}(s,t) = 2^{j}\Psi(2^{j}s-k_{1})\Psi(2^{j}t-k_{2})$
 $\Psi_{j,\kappa_{1},\kappa_{2}}^{HH}(s,t) = 2^{j}\Psi(2^{j}s-k_{1})\Psi(2^{j}t-k_{2})$

Locality:

At a scale j and a shift (K_1,K_2) The Y_{j,K_1,K_2}^{LH} , Y_{j,K_1,K_2}^{HL} , Y_{j,K_1,K_2}^{HH} , Y_{j,K_1,K_2}^{HH} , are centered at $x(K_12^{-j},K_22^{-j})$ and concentrated on a square of width $\sim 2^{-j}$. K_22^{-j} , K_32^{-j} .

As a general rule, wavelet coefficients will be small it the image is smooth over this region, and will be "active" if there is an edge or texture in this region.